

Micro-Power Inverting DC/DC Controller

■ FEATURES

- 2.4V to 7V Input Voltage Operation.
- Adjustable Output Voltage up to -40V.
- Low Quiescent Current at 80 μ A.
- Pulse Frequency Modulation Maintains High Efficiency (87%).
- 70KHz to 160KHz Switching Frequency.
- Power-Saving Shutdown Mode (0.7 μ A Typical).
- High Efficiency with Low Cost External PNP Bipolar Transistor.

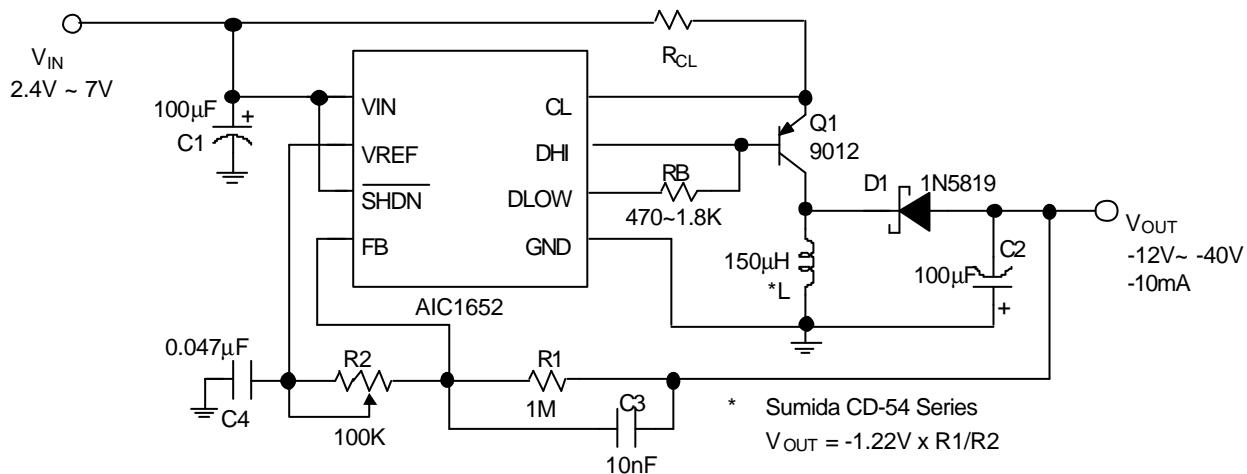
■ APPLICATIONS

- Negative LCD Contrast Bias for
 1. Notebook & Palmtop Computers.
 2. Pen-Based Data System.
 3. Portable Data Collection Terminals.
 4. Personal Digital Assistants.
- Negative Voltage Supply.

■ DESCRIPTION

The AIC1652 is a high performance inverting DC/DC controller, designed to drive an external power switch to generate programmable negative voltages. In the particularly suitable LCD bias contrast application, efficiency of 87% can be achieved with low cost PNP bipolar transistor drivers. Output voltage can be scaled to -40V or greater by two external resistors. A pulse frequency modulation scheme is employed to maintain high efficiency conversion under wide input voltage range. Quiescent current is about 80 μ A and can be reduced to 0.7 μ A in shutdown mode. Switching frequency being around 70KHz to 160KHz range, small size switching components are ideal for battery powered portable equipments, like notebook and palmtop computers.

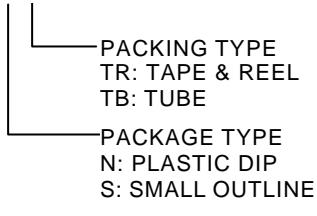
■ TYPICAL APPLICATION CIRCUIT



Negative LCD Contrast Bias Power Supply

■ ORDERING INFORMATION

AIC1652CXXX

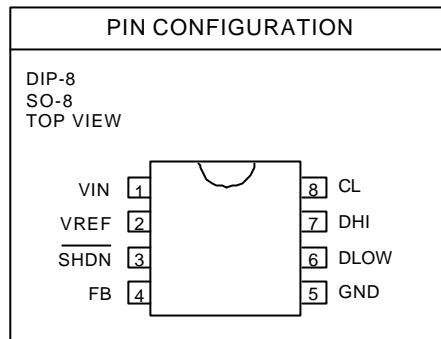


EX: AIC1652CSTR

→ in SO-8 Package & Tape & Reel

Packing Type

(CN is not available in TR packing type.)



■ ABSOLUTE MAXIMUM RATINGS

Supply Voltage 7V

SHDN Voltage 7V

Operation Temperature Range 0°C~70°C

Storage Temperature Range -65°C~ 150°C

■ TEST CIRCUIT

Refer to Typical Application Circuit.

■ ELECTRICAL CHARACTERISTICS ($V_{IN}=5V$, $T_a=25^\circ C$, unless otherwise specified.)

| PARAMETER | TEST CONDITIONS | MIN. | TYP. | MAX. | UNIT |
|--------------------------|-------------------------|------|------|------|----------|
| Input Voltage | | 2.4 | 7 | | V |
| Switch Off Current | $V_{FB}=-50mV$ | | 80 | 150 | μA |
| V_{REF} Voltage | $I_{SOURCE} = 250\mu A$ | 1.16 | 1.22 | 1.28 | V |
| V_{REF} Source Current | | 250 | | | μA |
| DLOW "ON Resistance" | | | 5 | | Ω |
| DHI "ON Resistance" | | | 7 | | Ω |
| CL Threshold | | | 70 | | mV |
| Shutdown Threshold | | 0.8 | 1.5 | 2.4 | V |
| Shutdown Mode Current | $V_{SHDN}=0V$ | | 0.7 | 2 | μA |

■ TYPICAL PERFORMANCE CHARACTERISTICS

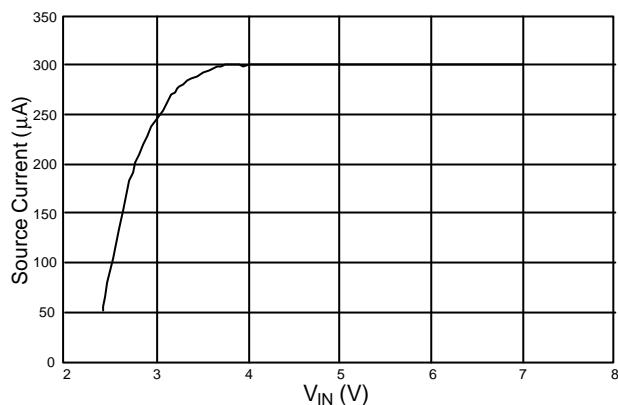


Fig. 1 V_{REF} Source Current vs. V_{IN}

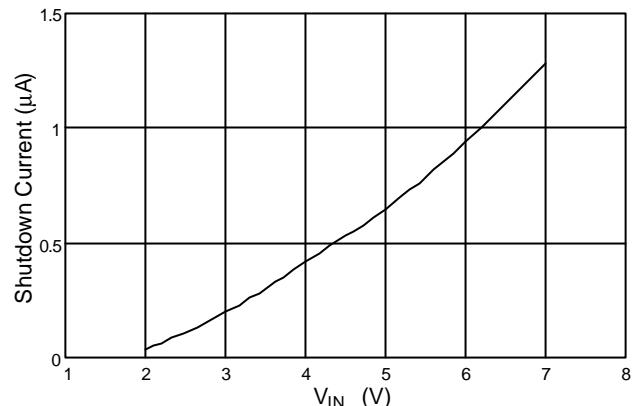


Fig. 2 Shutdown Current vs. V_{IN}

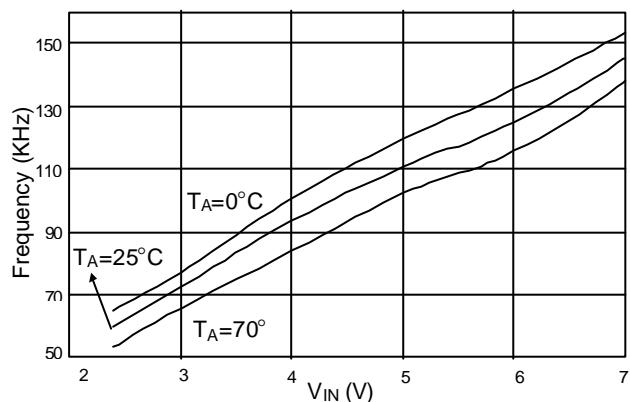


Fig. 3 Frequency vs. V_{IN} Voltage

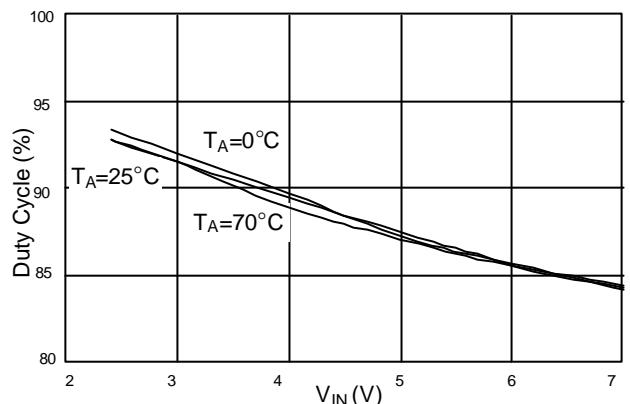
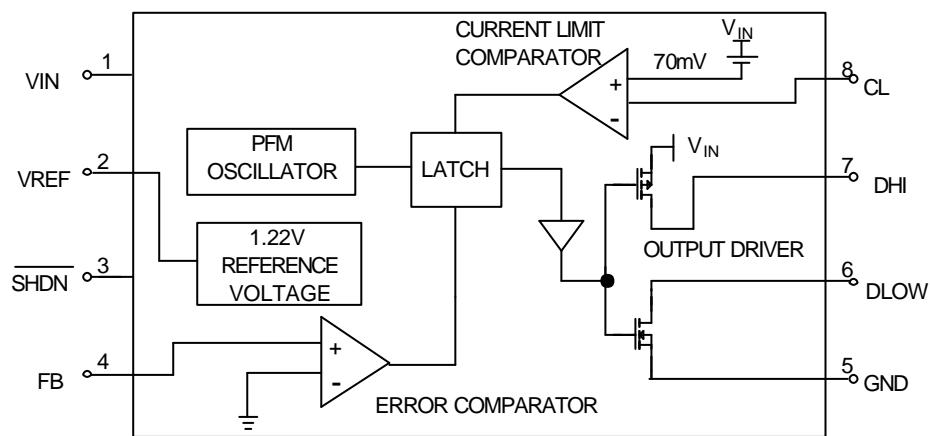


Fig. 4 Duty Cycle vs. V_{IN} Voltage

■ BLOCK DIAGRAM



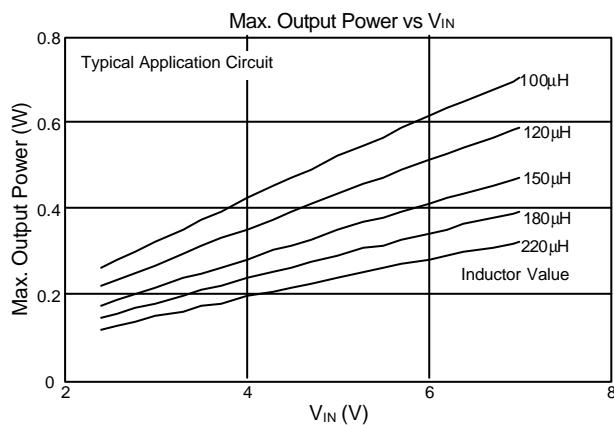
■ PIN DESCRIPTIONS

- PIN 1: VIN - Input supply voltage (2.4V~7V)
- PIN 2: VREF - Reference output (1.22V). By-pass with a $0.047\mu F$ capacitor to GND. Sourcing capability is guaranteed to be greater than $250\mu A$.
- PIN 3: SHDN - Logic input to shutdown the chip.
 $>1.5V$ = normal operation,
GND = shutdown
In shutdown mode DLOW and DHI pins are at high level.
- PIN 4: FB - Feedback signal input to sense ground. Connecting a resistor R1 to VOUT and a resistor R2 to VREF pin yields the output voltage:
 $V_{OUT} = -(R1/R2) \times V_{REF}$
- PIN 5: GND - Power ground.

PIN 6: DLOW - Driver sinking output. Connected to DHI when using an external P-channel MOSFET. When using an external PNP bipolar transistor, connect a resistor RB from this pin to DHI. RB value depends on V_{IN} , inductor and PNP bipolar transistor. By adjusting the RB value, efficiency can be optimized.

- PIN 7: DHI - Driver sourcing output. Connect to gate of the external P-channel MOSFET or base of the PNP bipolar transistor.
- PIN 8: CL - Current-limit input. This pin clamps the switch peak current to prevent over-current damage to the external switch.

■ APPLICATION INFORMATIONS



The typical application circuit generates an adjustable negative voltage for contrast bias of LCD displays. Efficiency and output power can be optimized by using appropriate inductor and switch. The following formulas provide a guideline for determining the optimal component values:

$$L = (11.1 - 0.15 \times V_{IN}) \times \frac{V_{IN}}{|I_{OUT}| \times |V_{OUT}|}$$

$$\text{PNP : } |V_{CEO}| > V_{IN} + |V_{OUT}|$$

$$|I_{C,MAX}| \geq 200 \times \frac{|I_{OUT}|}{V_{IN}}$$

$$|V_{CE}| < 0.4V \text{ at } I_C = 200 \times \frac{|I_{OUT}|}{V_{IN}}$$

and $b = 10$

$$RB \approx 3 \times L \times (V_{IN} - 0.8)$$

where, $V_{IN}(V)$, $V_{OUT}(V)$, $I_{OUT}(A)$, $L(\mu H)$, $RB(\Omega)$

■ APPLICATION CIRCUIT (Refer to TYPICAL APPLICATION CIRCUIT)

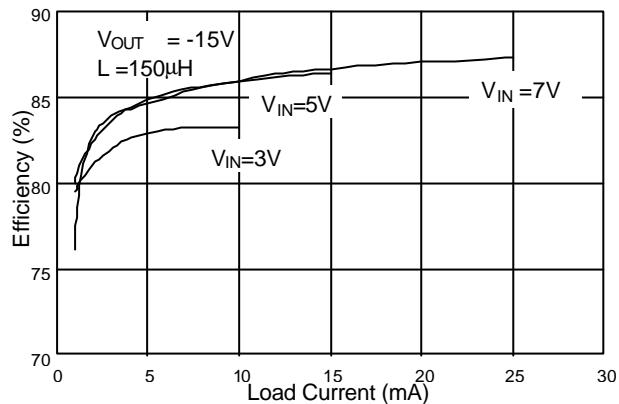


Fig. 5 Efficiency vs. Load Current

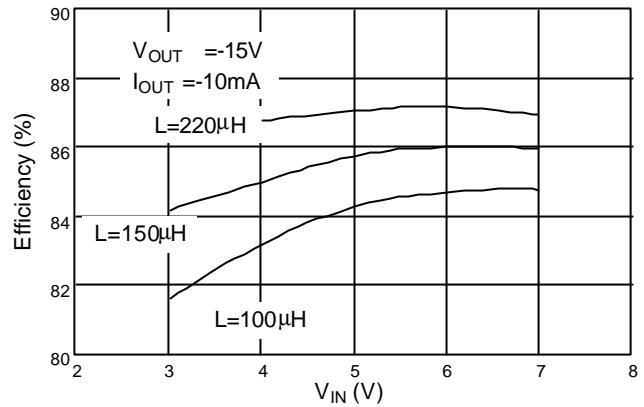


Fig. 6 Efficiency vs. V_{IN}

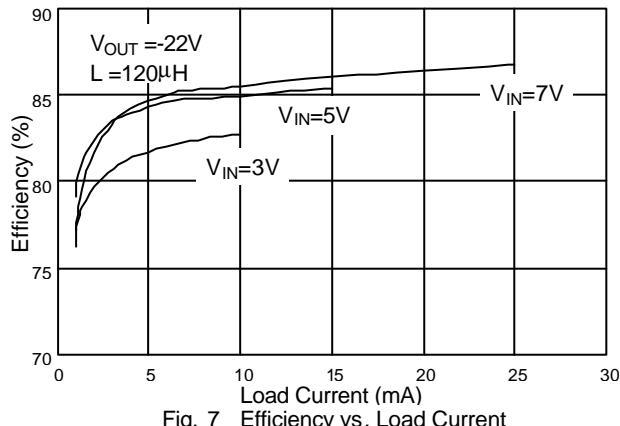


Fig. 7 Efficiency vs. Load Current

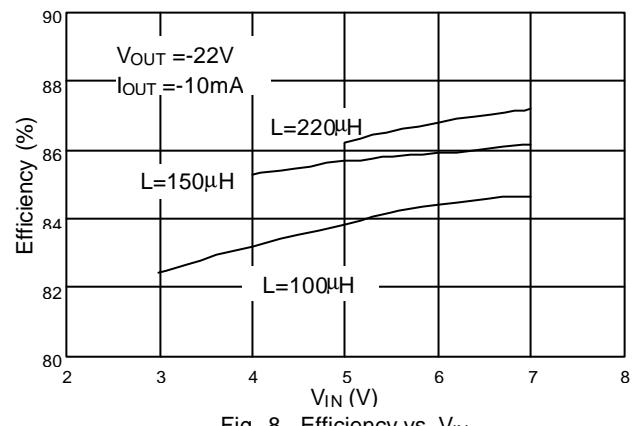


Fig. 8 Efficiency vs. V_{IN}

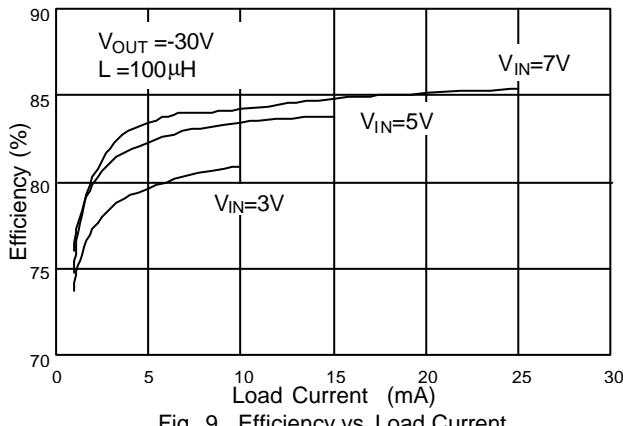


Fig. 9 Efficiency vs. Load Current

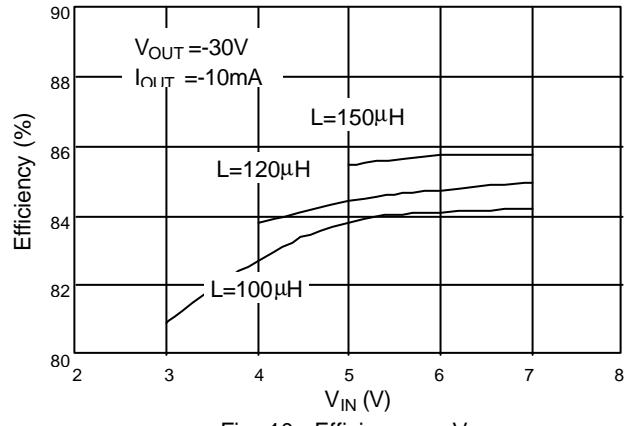
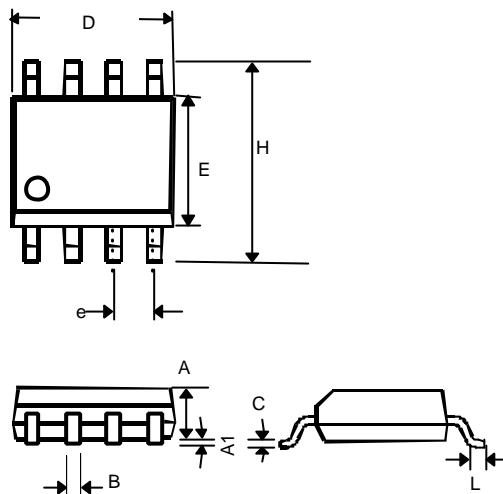


Fig. 10 Efficiency vs. V_{IN}

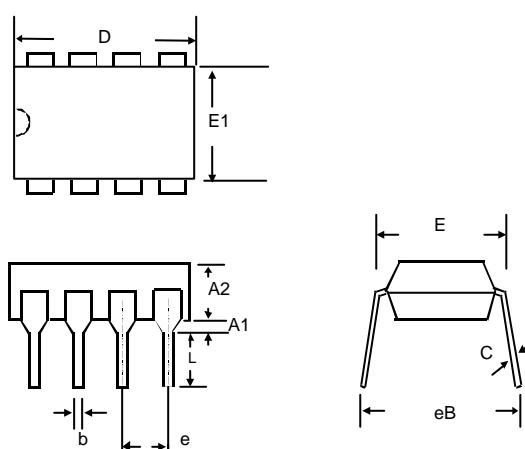
■ PHYSICAL DIMENSIONS

- 8 LEAD PLASTIC SO (unit: mm)



| SYMBOL | MIN | MAX |
|--------|-----------|------|
| A | 1.35 | 1.75 |
| A1 | 0.10 | 0.25 |
| B | 0.33 | 0.51 |
| C | 0.19 | 0.25 |
| D | 4.80 | 5.00 |
| E | 3.80 | 4.00 |
| e | 1.27(TYP) | |
| H | 5.80 | 6.20 |
| L | 0.40 | 1.27 |

- 8 LEAD PLASTIC DIP (unit: mm)



| SYMBOL | MIN | MAX |
|--------|------------|-------|
| A1 | 0.381 | — |
| A2 | 2.92 | 4.96 |
| b | 0.35 | 0.56 |
| C | 0.20 | 0.36 |
| D | 9.01 | 10.16 |
| E | 7.62 | 8.26 |
| E1 | 6.09 | 7.12 |
| e | 2.54 (TYP) | |
| eB | — | 10.92 |
| L | 2.92 | 3.81 |